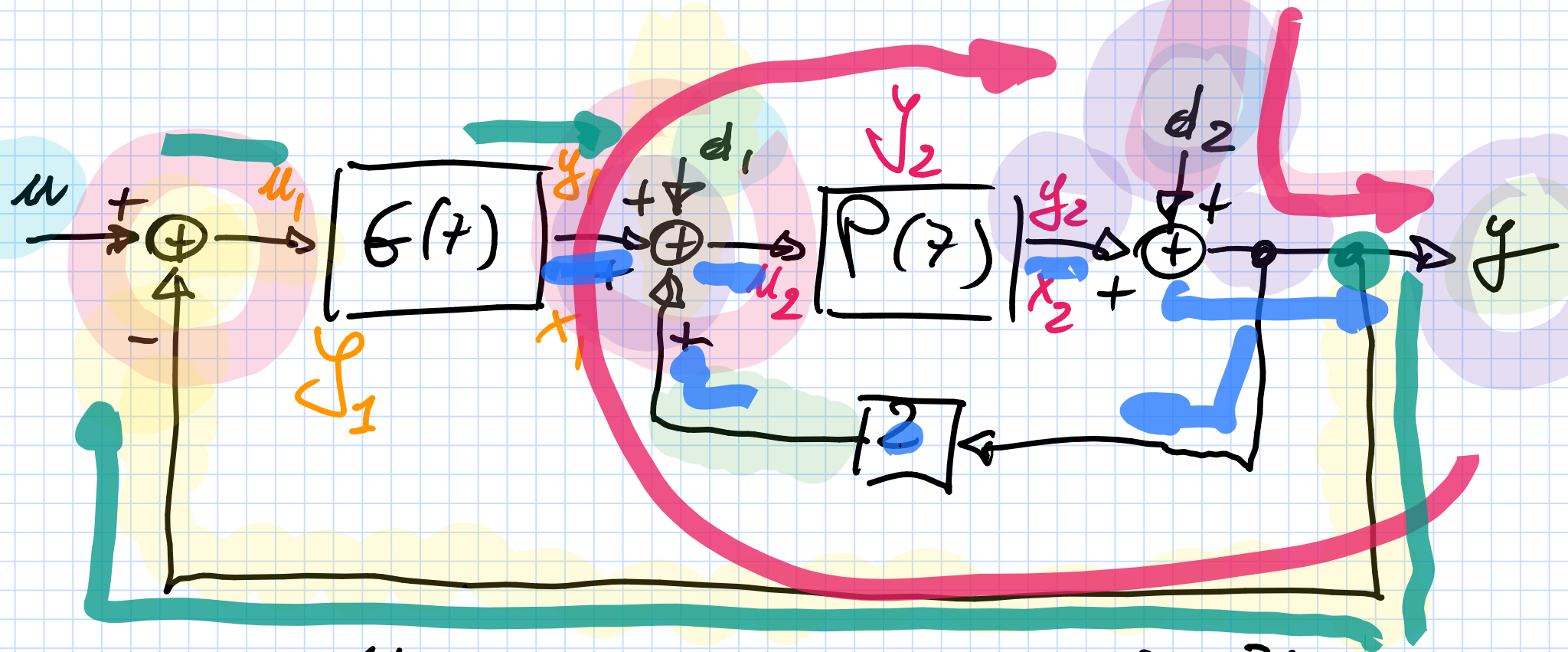


È scritto su  
manipolazione  
di schemi a  
blocchi

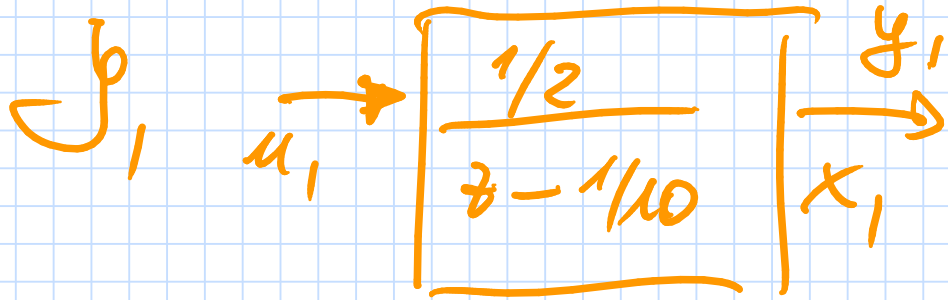
Resoluz. in equazioni di stato e manipolazione di  
 sistemi a blocchi



$$G(z) = \frac{1/2}{z - 1/10}$$

$$P(z) = \frac{z + 2/5}{z + 1/2}$$

① Determinare una realizzaione in ep. di stato



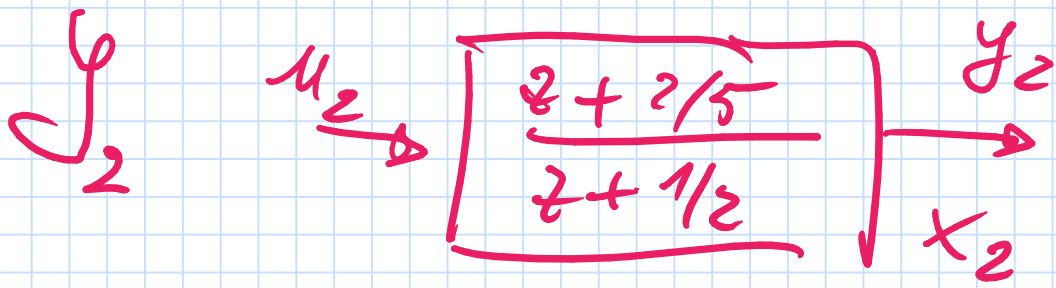
$$FdT \quad u=1 \\ u=0$$

$$G(z) = \frac{1/2}{z - 1/10}$$

$$T(z) = \frac{cb}{z - a}$$

$$\begin{cases} a = +\frac{1}{10} \\ cb = \frac{1}{2} \end{cases}$$

$$\begin{cases} x_1(k+1) = \frac{1}{10} x_1(k) + \frac{1}{2} u_1(k) \\ y_1(k) = x_1(k) \end{cases}$$



FdT  $\mu=1$   
 $\mu=1$

$$P(z) = \frac{z + 2/5}{z + 1/2}$$

$$T(z) = d + \frac{cb}{z - a}$$

$z + \frac{2}{5}$	$z + \frac{1}{2}$
$-z - \frac{1}{2}$	$1$
$1 - \frac{1}{10}$	

$$= 1 - \frac{1/10}{z + 1/2}$$

$$\begin{cases} x_2(k+1) = -\frac{1}{2}x_2(k) + \frac{1}{10}u_2(k) \\ y_2(k) = -x_2(k) + u_2(k) \end{cases}$$

$$\begin{cases} d = +1 \\ a = -\frac{1}{2} \\ cb = -\frac{1}{10} \end{cases}$$

$$x_1(k+1) = -\frac{1}{10}x_1(k) + \frac{1}{2}u_1(k)$$

$$x_2(k+1) = -\frac{1}{2}x_2(k) + \frac{1}{10}u_2(k)$$

$$y_1(k) = x_1(k)$$

$$y_2(k) = -x_2(k) + u_2(k)$$

$$y(k) = d_2(k) + y_2(k)$$

con blocchi e segnali generati

$$u_1(k) = u(k) - y(k)$$

$$u_2(k) = y_1(k) + d_1(k) + 2y(k)$$

$$\begin{cases} x_1(k+1) = \frac{1}{10} x_1(k) + \frac{1}{2} u(k) - \frac{1}{2} y(k) \\ x_2(k+1) = -\frac{1}{2} x_2(k) + \frac{1}{10} \begin{bmatrix} -x_1(k) + 2x_2(k) + \\ -d_1(k) - 2d_2(k) \end{bmatrix} \\ y(k) = -x_1(k) + x_2(k) - d_1(k) - d_2(k) \end{cases}$$

Per caso  $\rightarrow$  terminare la sostituzione ...

2 variabili di stato  $\rightarrow$  ordine 2

3 ingressi ( $u, d_1, d_2$ )

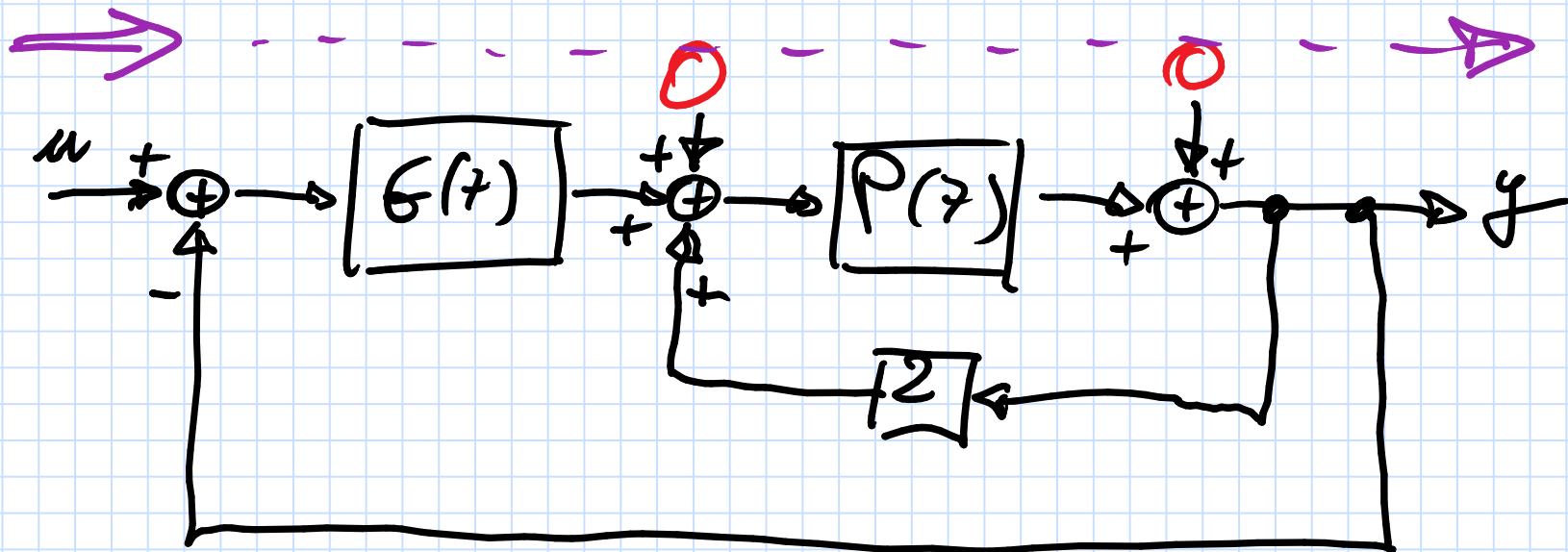
1 uscita  $y$

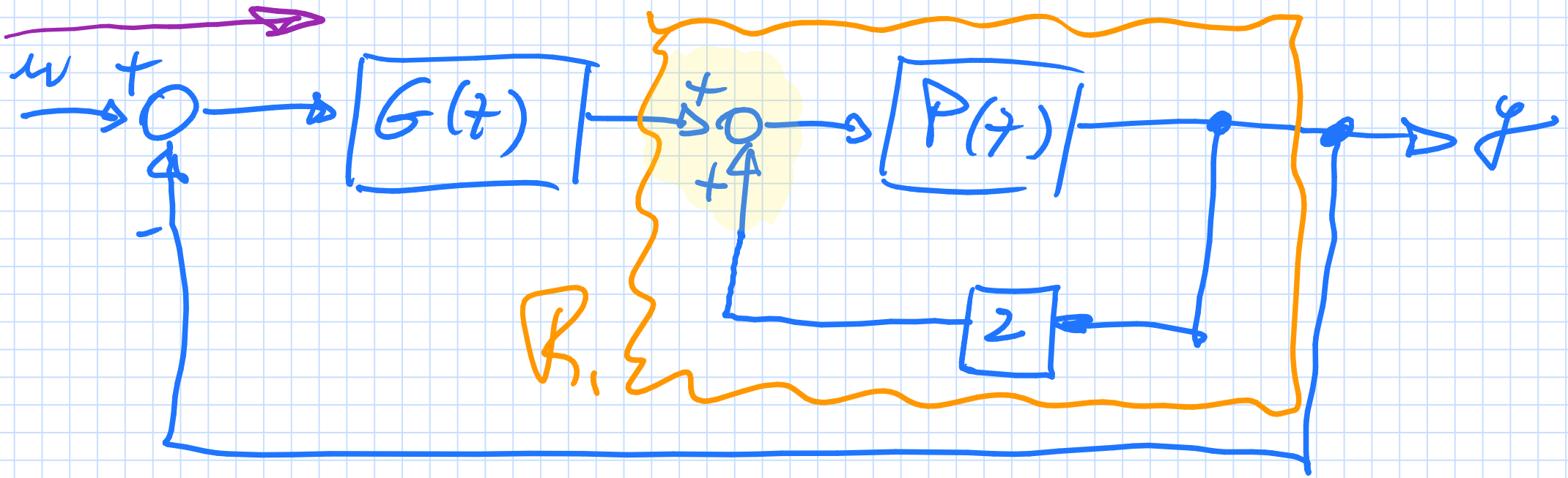
① Determinare la FdT tra  $u(k)$ ,  $d_1(k)$ ,  $d_2(k)$  e l'uscita  $y(k)$

Sovrapposizione degli effetti  $\rightarrow$

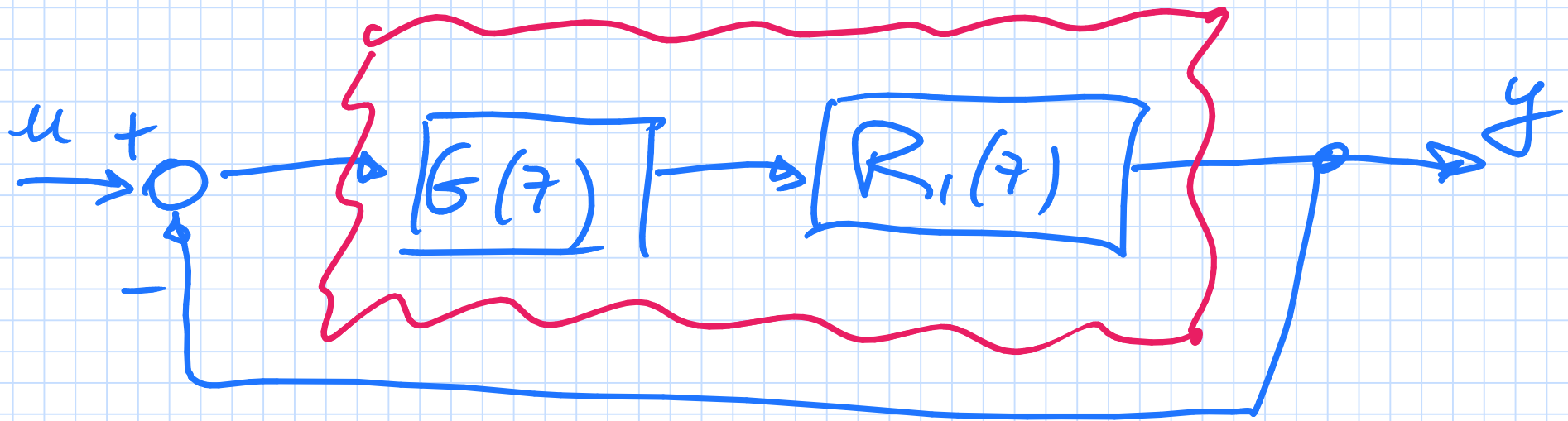
- farlo 2 input e due
- ridisegnare lo schema
- calcolare la FdT tra l'input rimasto ed  $y$

1° caso FdT tra  $u(k)$  e  $y(k)$





$$R_1(z) = \frac{P(z)}{1 - 2P(z)}$$





$$T_{u,y}(z) = \frac{G(z) \cdot R_1(z)}{1 + G(z) R_1(z)}$$

$$R_1(z) = \frac{P(z)}{1 - zP(z)}$$

$$= \frac{G(z) P(z)}{1 - 2P(z) + G(z) P(z)}$$

$$= \frac{1 - P(z) [2 - G(z)]}{1 - 2P(z) + G(z) P(z)}$$

$$= \frac{-0,5(z + 0,4)}{z^2 - 0,3z - 0,23}$$

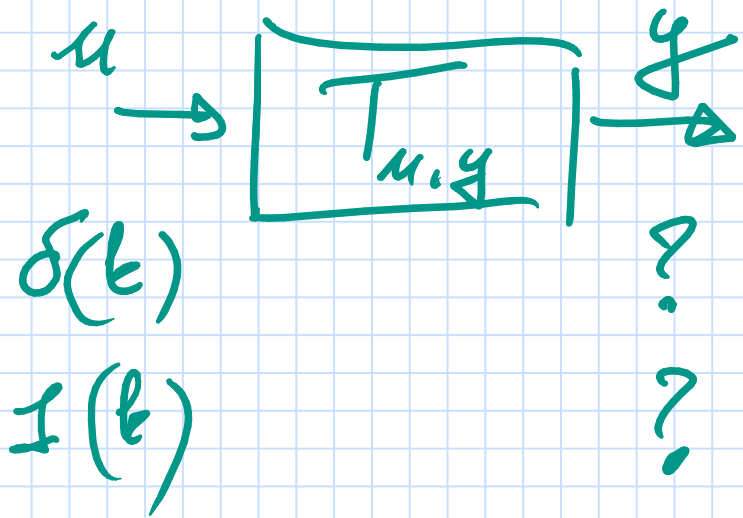
$$G(z) = \frac{1/2}{z - 1/10}$$

$$P(z) = \frac{z + 2/5}{z + 1/2}$$

$$m = 2$$

$$u = 1$$

$$T_{u,y}(z) = \frac{-0,5(z + 0,4)}{z^2 - 0,3z - 0,73}$$



Wir lösen alle Singular

$$y(0) = 0$$

$$y(1) = -\frac{1}{2}$$

$$\mathcal{Z}\{\delta(k)\} = 1$$

$$Y(z) = T_{u,y}(z) U(z)$$

$$\text{für } \delta(k) \rightarrow U(z) = 1$$

$$\text{für } I(k) \rightarrow U(z) = \frac{z}{z-1}$$

$$Y(z) = T_{u,y}(z) \cdot 1 = \frac{-0,5(z + 0,4)}{z^2 - 0,3z - 0,23}$$

Strof. presenza del moltiplicatore?

$$g(0) = \lim_{z \rightarrow +\infty} Y(z) = 0$$

$$g(1) = \lim_{z \rightarrow +\infty} z \left\{ Y(z) \right\} = \lim_{z \rightarrow +\infty} z \cdot Y(z) = -\frac{1}{2}$$

Risposta allo scalino

$$y_s(k)$$

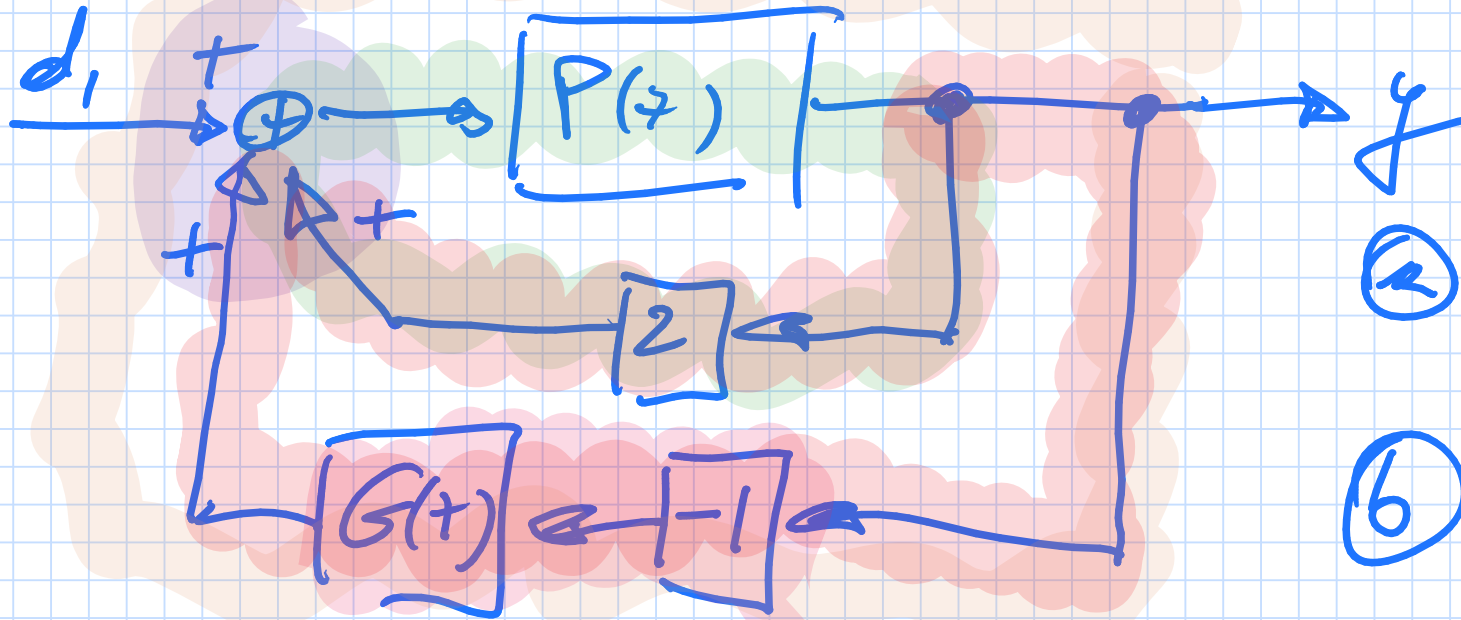
$$y_s(0) = 0$$

$$y_s(1) \neq 0$$

$T_{d_1, y}(z)$

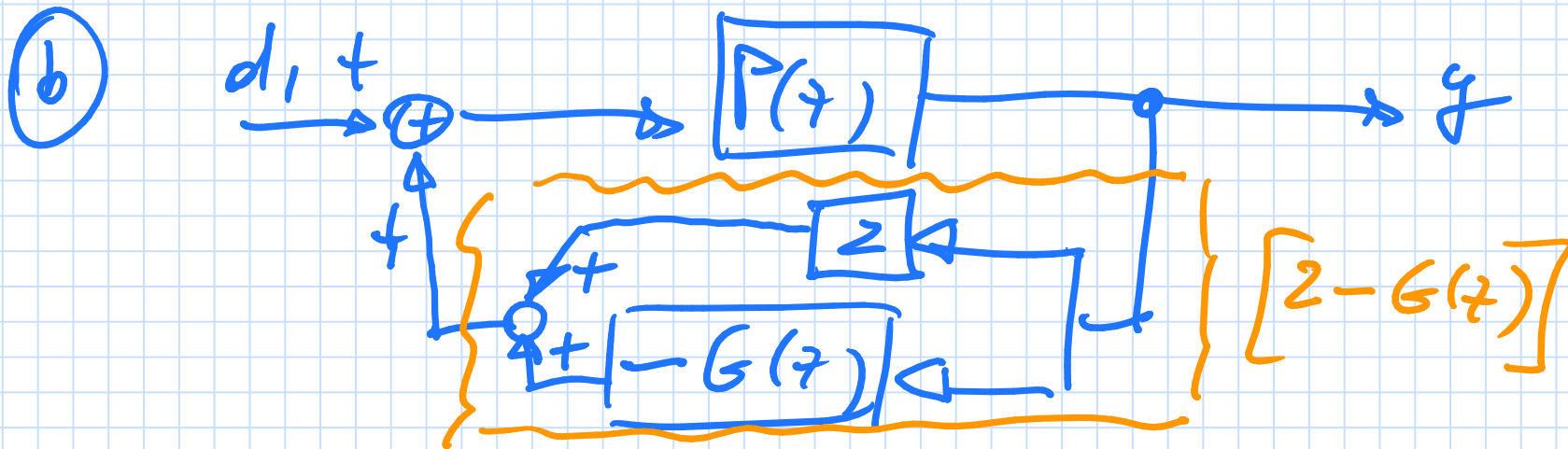
$$u(k) \equiv 0$$

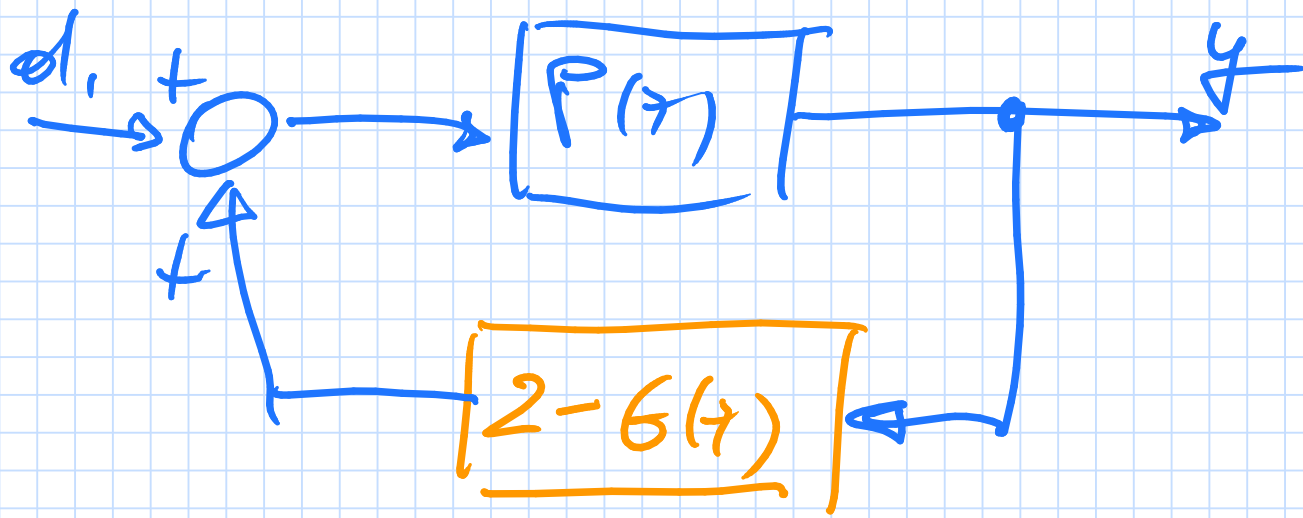
$$d_2(k) \equiv 0$$



(a) 2 retroazioni

(b) parallelo + retroazione



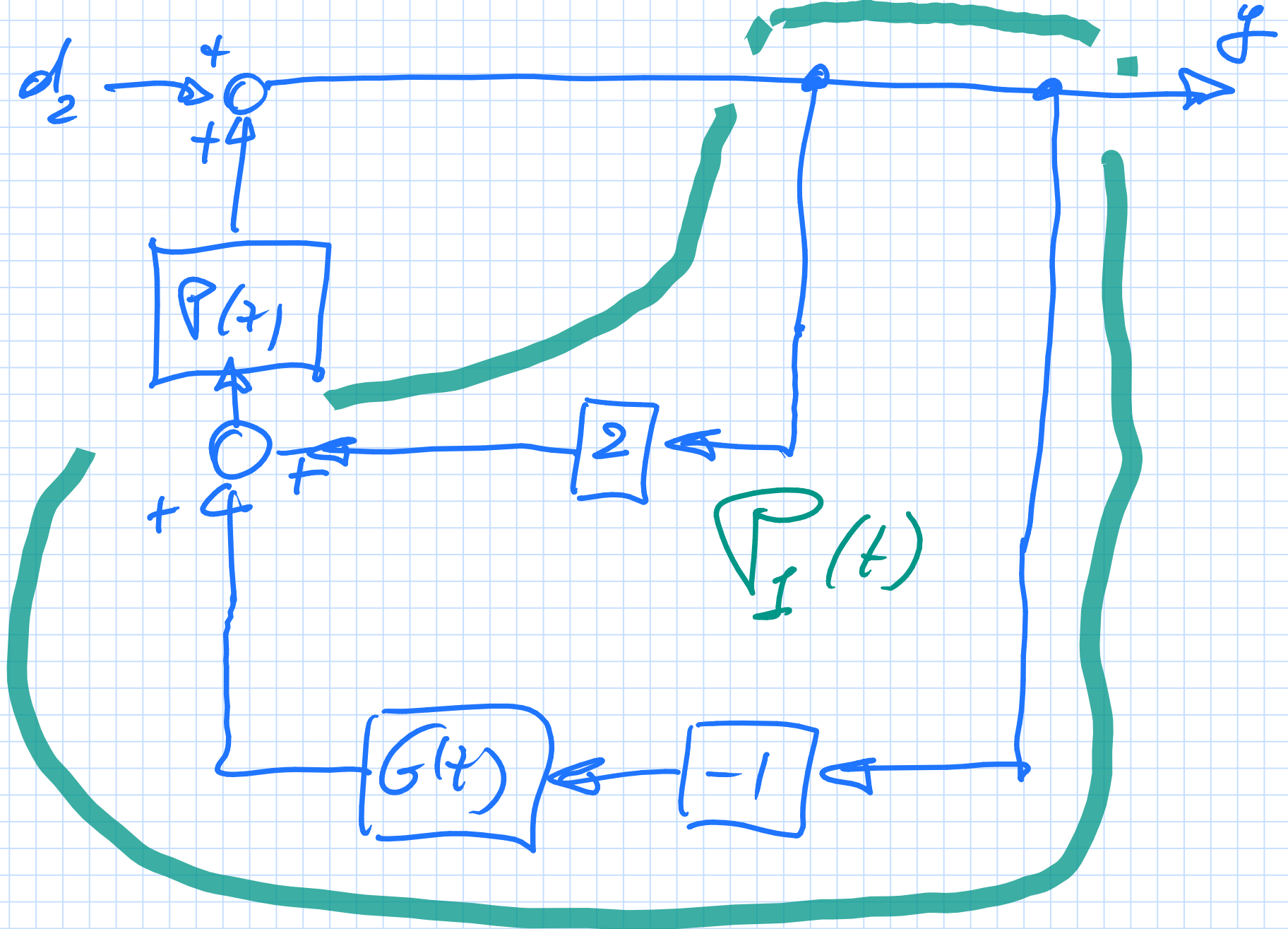


$$T_{d_i y}(s) = \frac{P(s)}{1 - [2 - G(s)] \cdot P(s)}$$

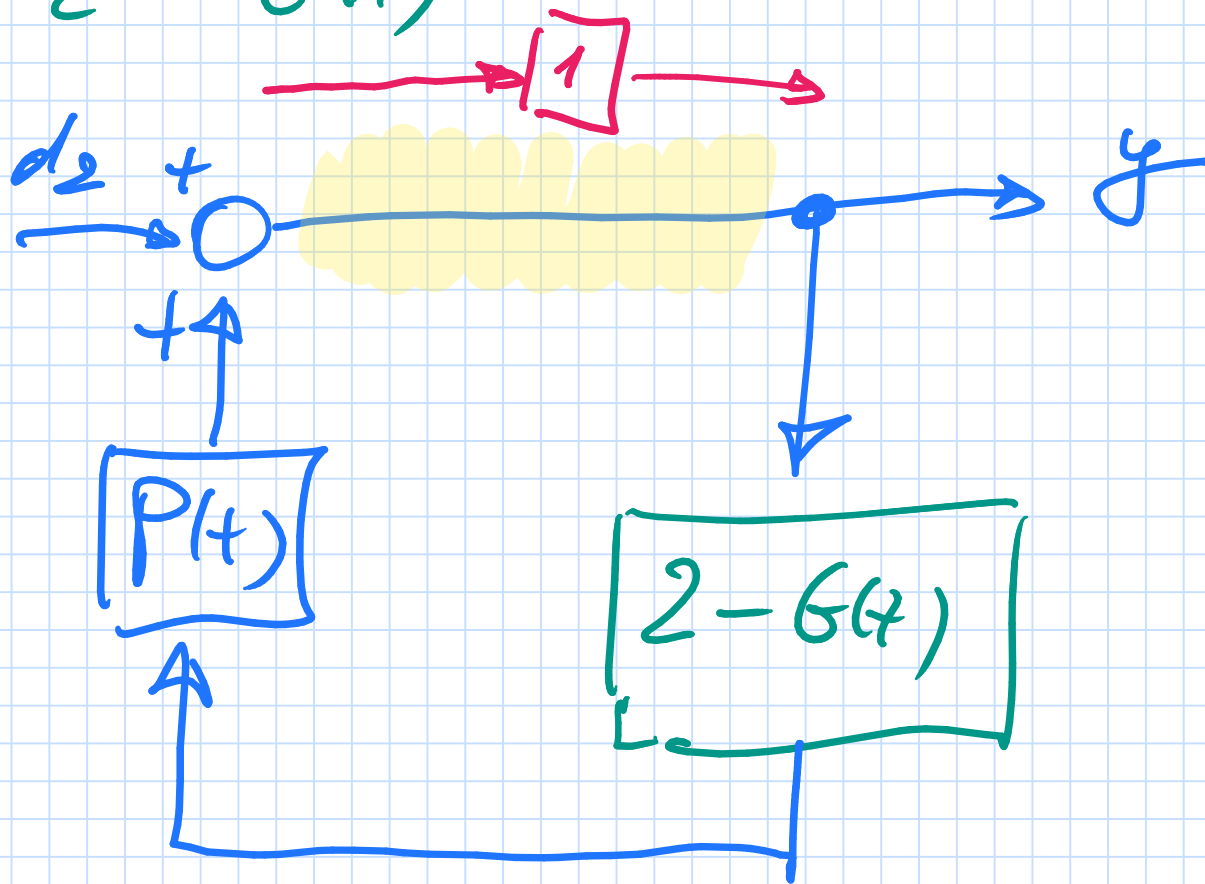
per ottenere l'espressione  
venne sostituite le espressioni  
di  $P(s)$  e  $G(s)$

$T_{d_2} y(t)$

$u(k) \equiv 0 \quad d_1 \equiv 0 \quad d_2$

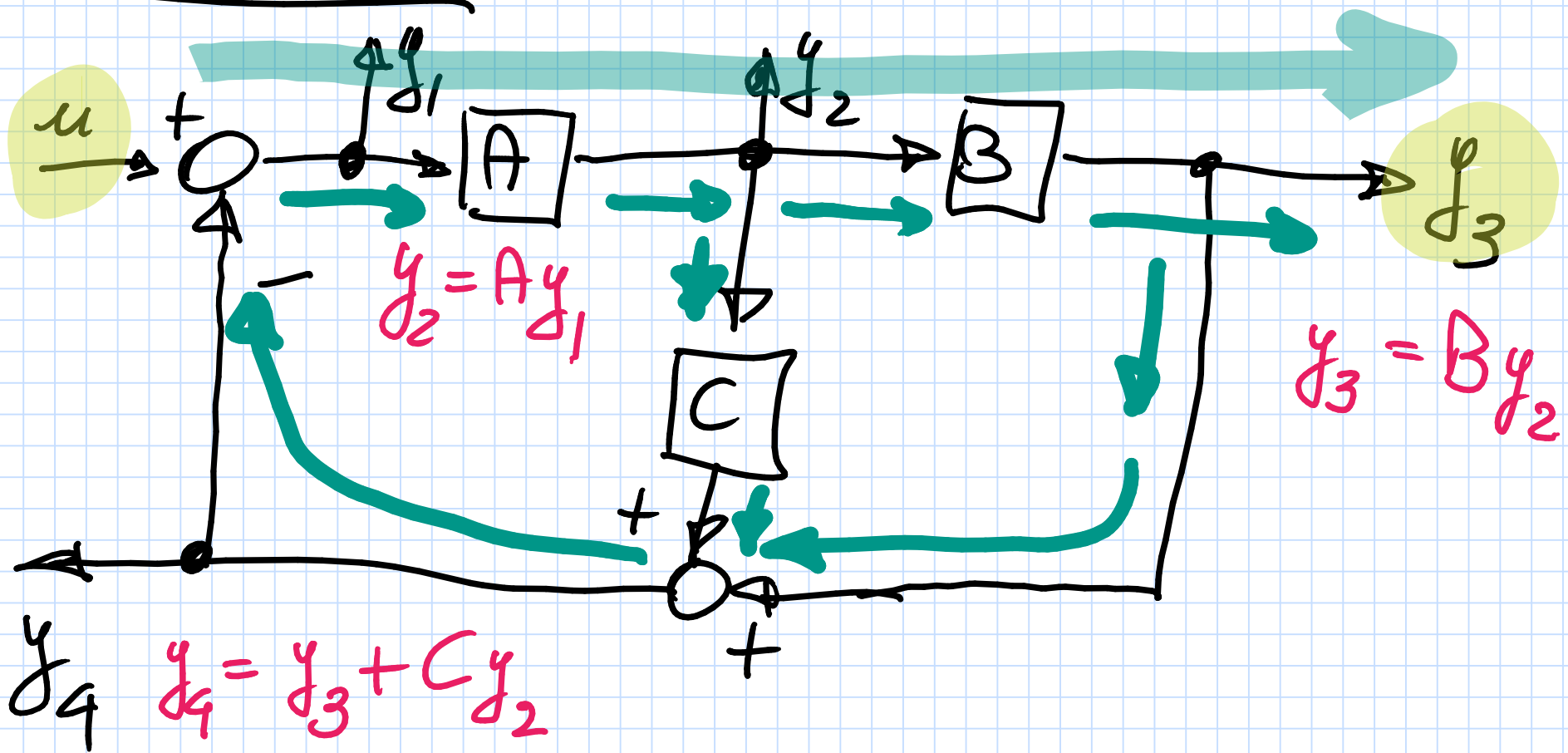


$$D_1(s) = 2 - G(s)$$



$$T_{d_2 y}(s) = \frac{1}{1 - P(s)[2 - G(s)]}$$

Es. 1 1/6/2005 / dato lo schema e blocchi



$y_4 = y_3 + Cy_2$

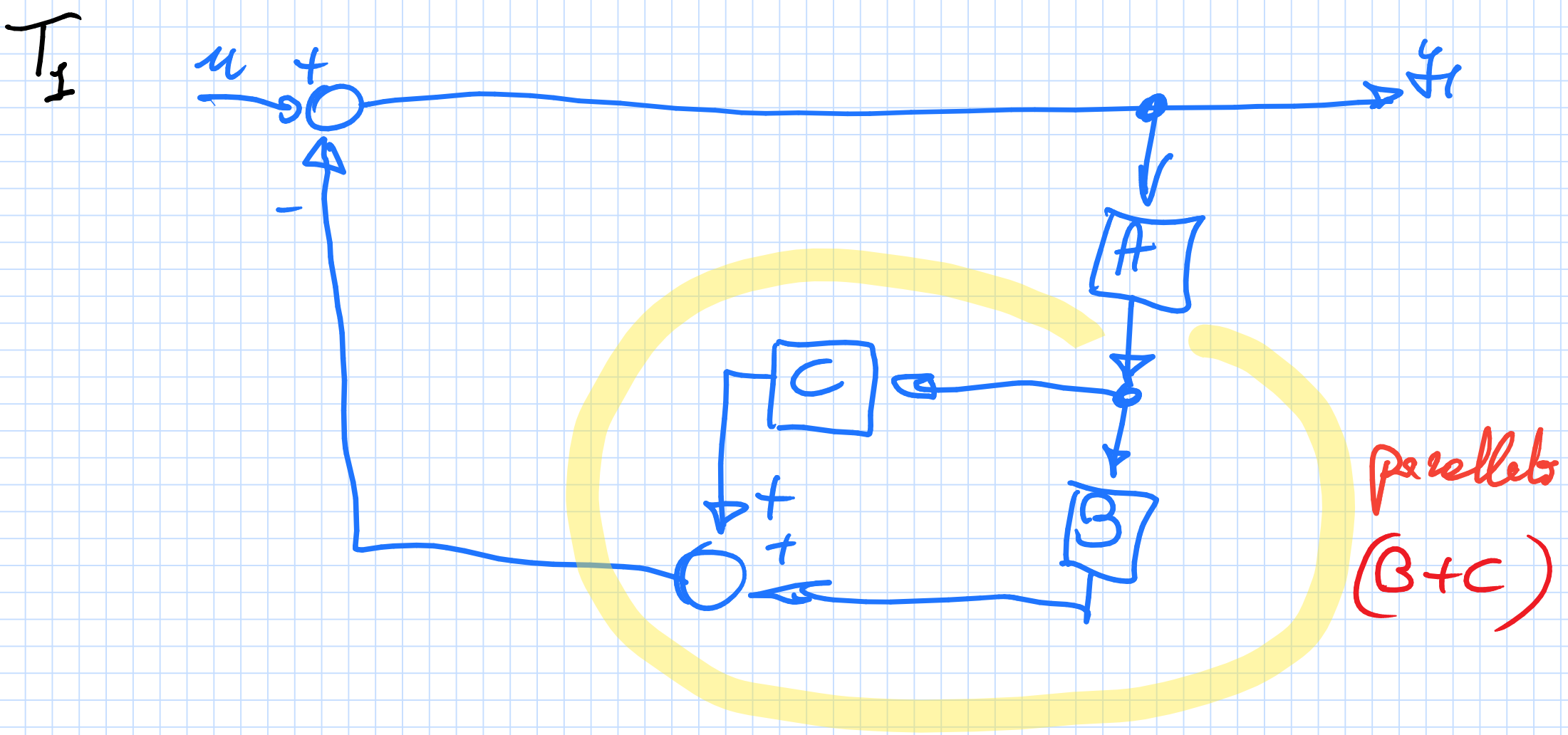
$T_{u, y_1} = ?$

$T_{u, y_2} = ?$

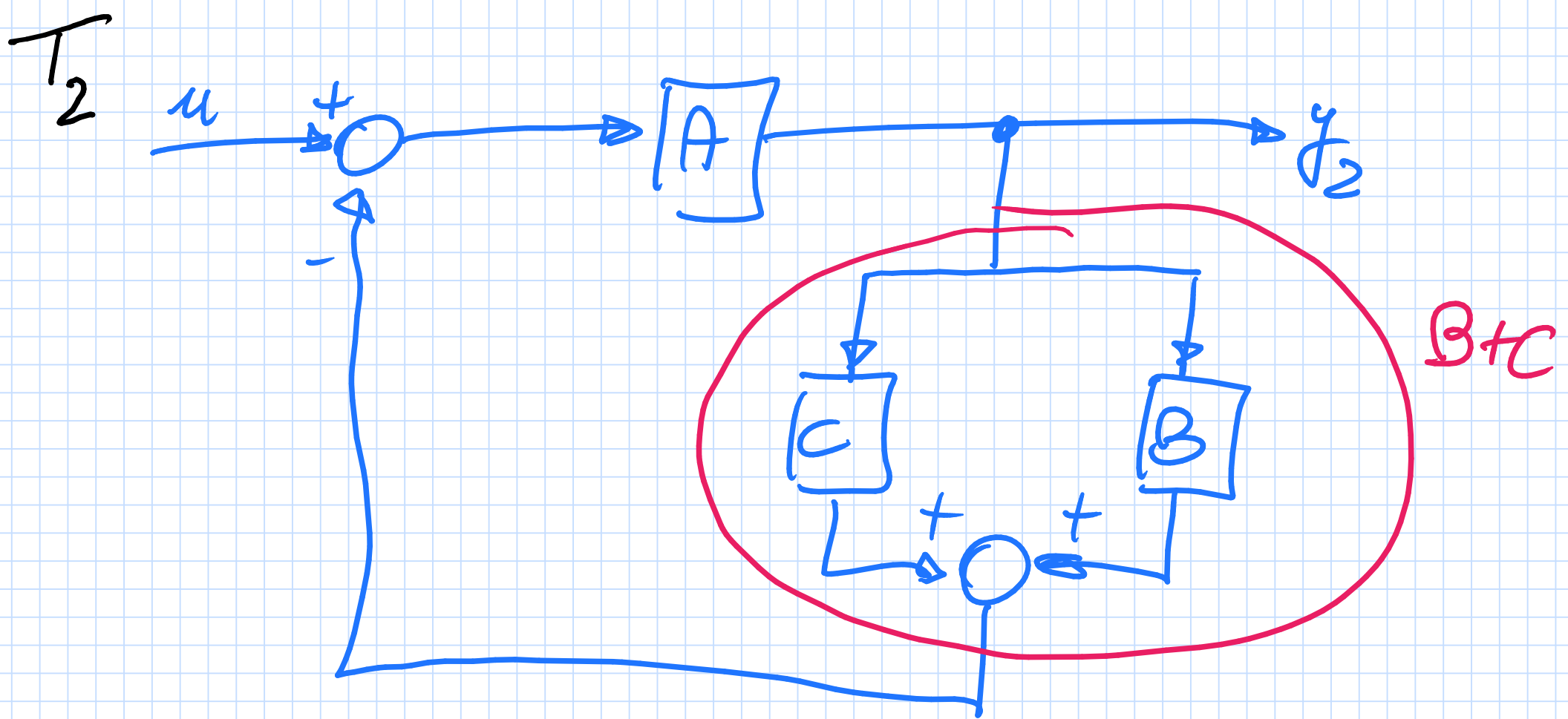
$T_{u, y_3} = ?$

$T_{u, y_4} = ?$



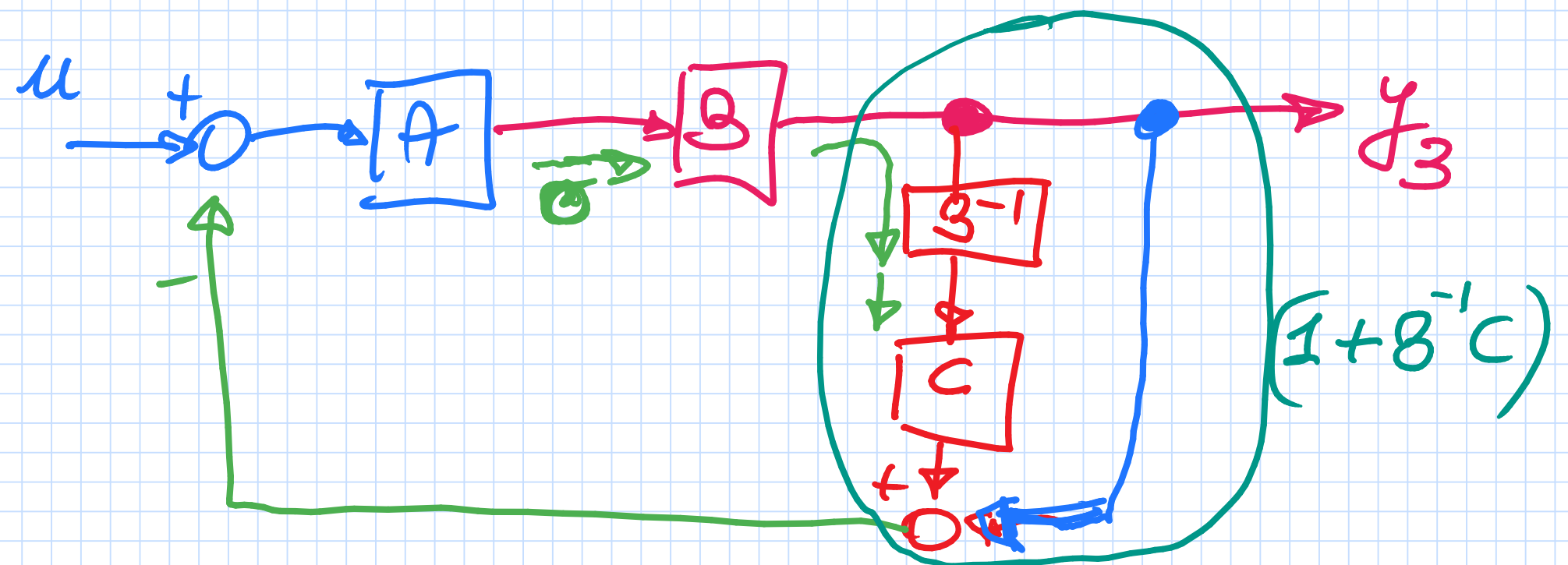
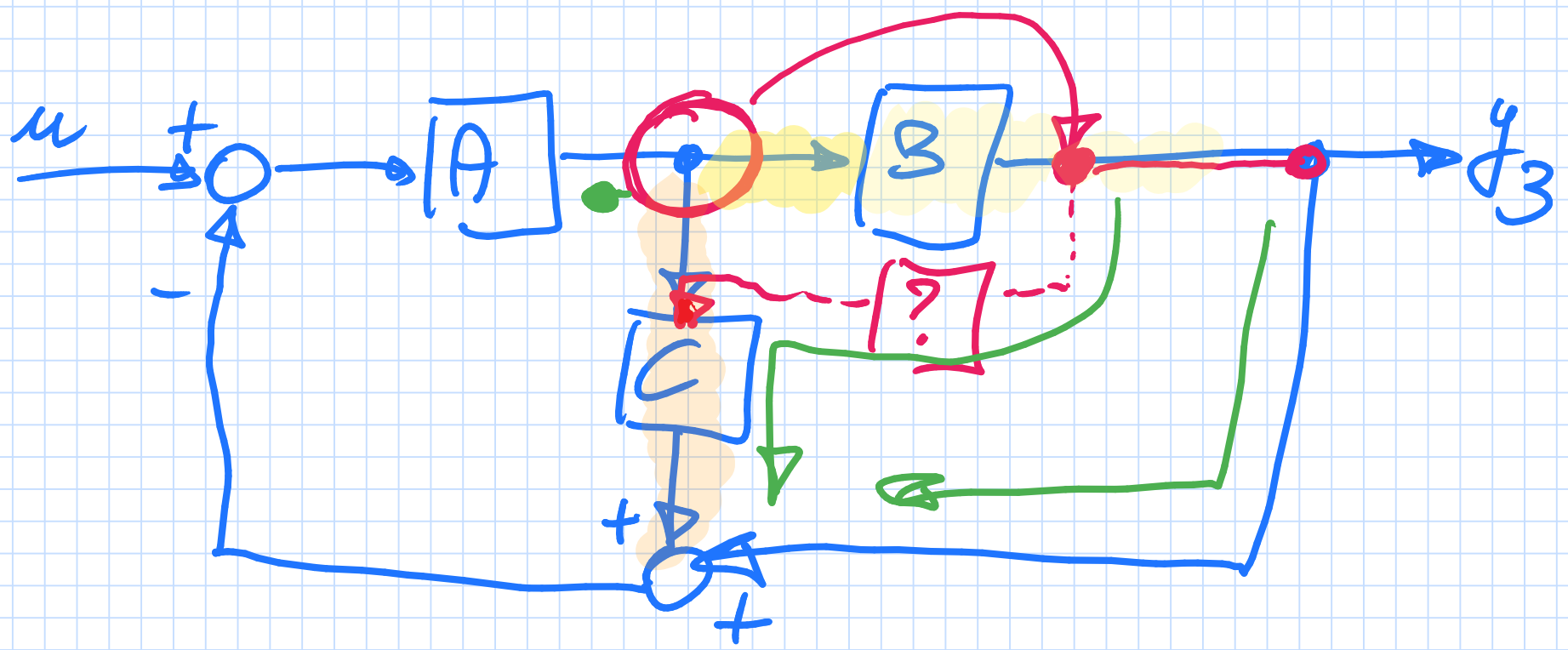


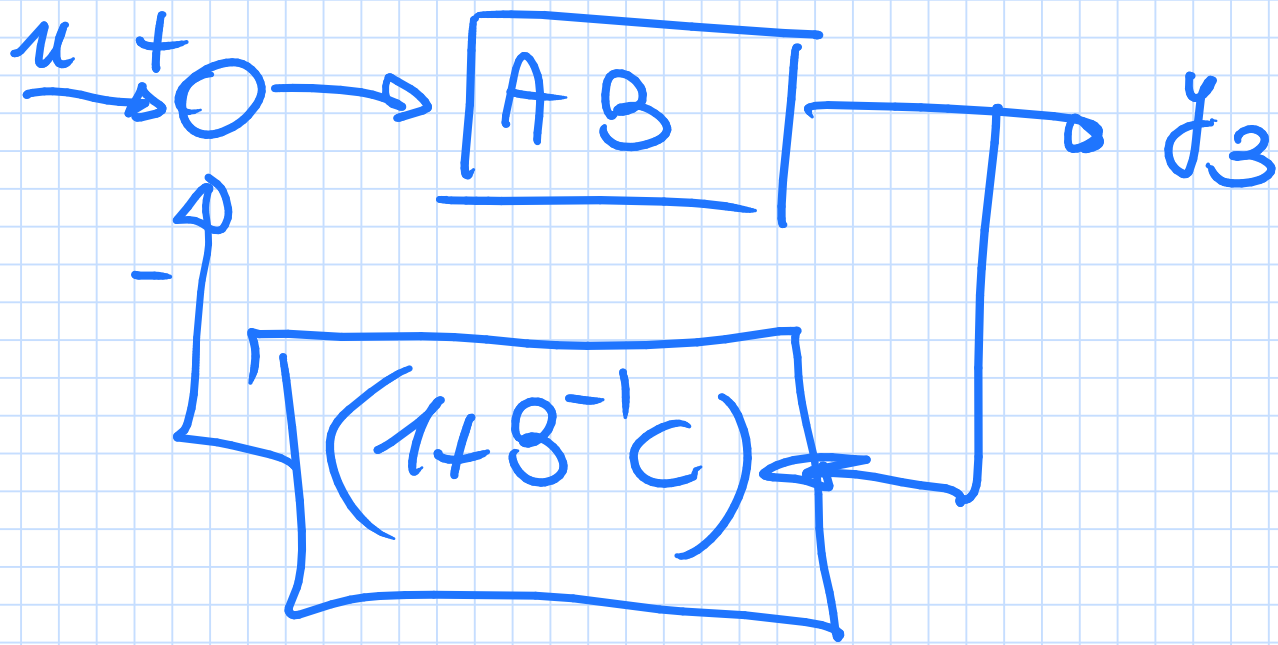
$$T_I = \frac{1}{1 + A(B+C)}$$



$$T_2 = \frac{A}{1 + A[B+C]} = A \cdot T_1$$

$T_3$





$$T_3 = \frac{AB}{1 + AB(1 + s^{-1}c)} = \frac{AB}{1 + \cancel{AB} \cancel{s^{-1}}(B+C)}$$

$$= \frac{AB}{1 + A(B+C)} = A \cdot B \cdot T_1$$

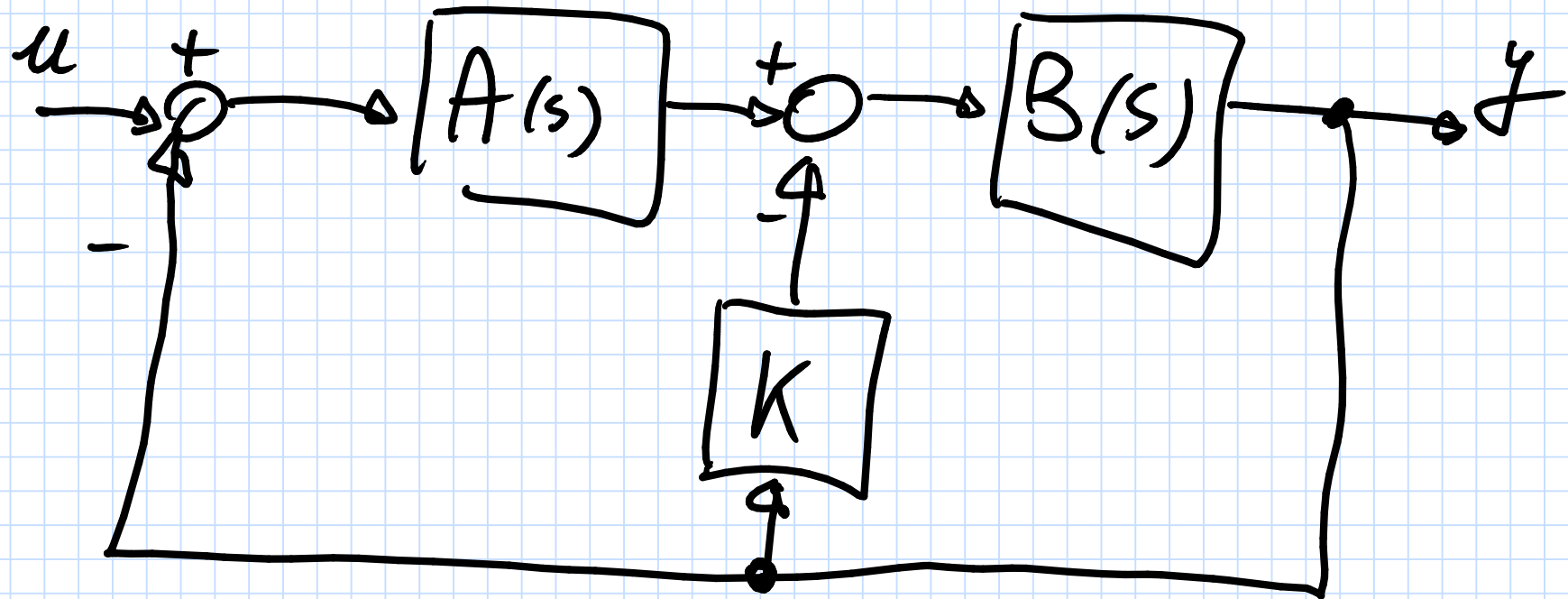
$T_4$ 

per case

$$T_4 = \frac{A(b+c)}{1 + A(b+c)}$$

$$y_4 = y_3 + C y_2$$

Per caso



$$A(s) = \frac{2}{s+1}$$

$$B(s) = \frac{1}{s-2}$$

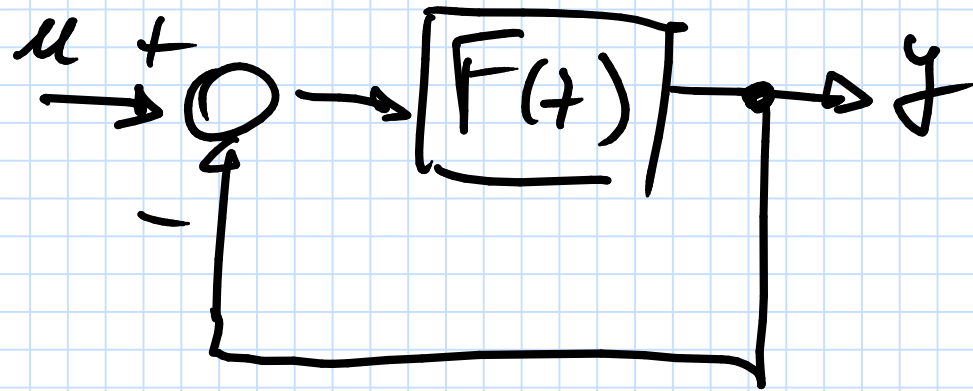
$$K \in \mathbb{R}$$

(•) realitt. in ep di stato

(•)  $T_{u,y}$

$$T_{u,y} = \frac{2}{s^2 + (k-1)s + k}$$

# Análisis de estabilidad



$$F(z) = \frac{z - a}{(z - 1)(z - \frac{1}{2})}$$

$$a \in \mathbb{R}$$

(a) estabilidad a cado de  $a \in \mathbb{R}$

(b) para  $a = +\frac{1}{2}$   $y(k) \xrightarrow{k \rightarrow +\infty} ?$

$$u(k) = 1(k) - 1(k-1) + 1(k-2) - 1(k-3)$$

Sol. (a) 
$$W(z) = \frac{F(z)}{1 + F(z)}$$

$$W(z) = \frac{(z-a)(z-1)(z-\frac{1}{2})}{1 + \frac{z-a}{(z-1)(z-\frac{1}{2})}} = \frac{z-a}{z^2 - \frac{1}{2}z + (\frac{1}{2}-a)}$$

$$P(z) = N(z) + D(z)$$

$$F(z) = \frac{N(z)}{D(z)}$$

$$P(z) = z^2 - \frac{1}{2}z + (\frac{1}{2}-a) = 0 \quad a \in \mathbb{K}$$

$$| \cdot | \leq 1$$

$$z = \frac{w+1}{w-1} \Rightarrow \left( \frac{w+1}{w-1} \right)^2 - \frac{1}{2} \left( \frac{w+1}{w-1} \right)^2 + \left( \frac{1}{2}-a \right) = 0$$

~~$(w-1)^2$~~

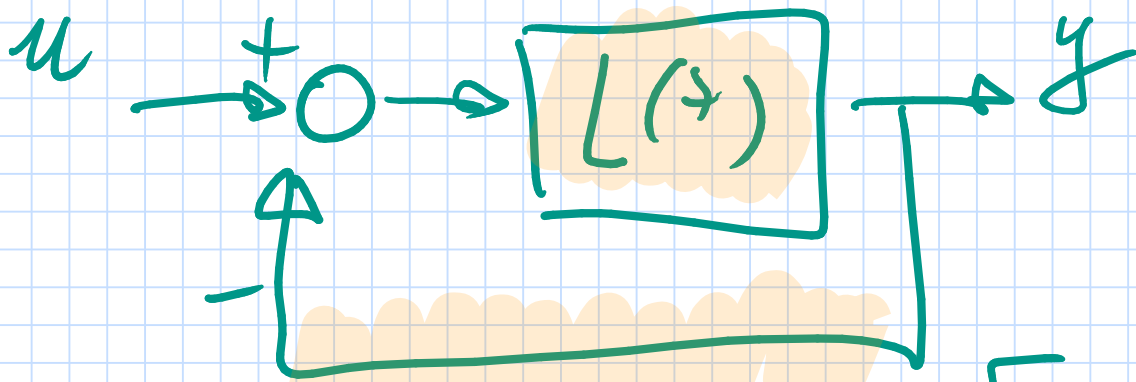


$$(1-a)w^2 + (2a+1)w + (2-a) = 0$$

$q(w)$

Criterio di Routh - Hurwitz

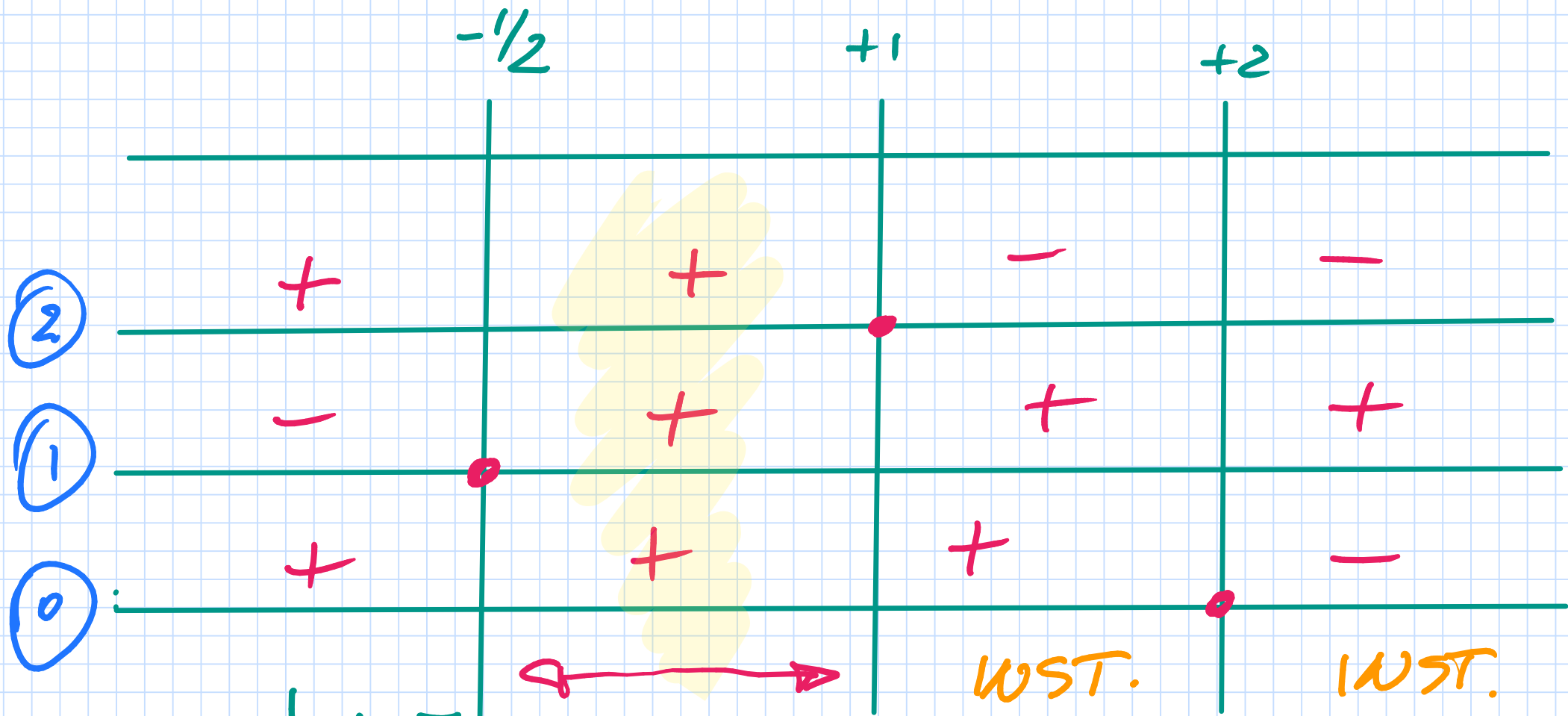
②	$(1-a)$	$(2-a)$	$1-a \geq 0$	$a \leq 1$
①	$(2a+1)$		$2a+1 \geq 0$	$a \geq -\frac{1}{2}$
①	$(2-a)$		$2-a \geq 0$	$a \leq 2$



$$L(z) = \frac{N(z)}{D(z)}$$

$$F(z) = \frac{L(z)}{1 + L(z)}$$

$$F(z) = \frac{\frac{N(z)}{D(z)}}{1 + \frac{N(z)}{D(z)}} = \frac{N(z)}{N(z) + D(z)}$$



$\alpha < -\frac{1}{2}$  WST

$\alpha = -\frac{1}{2}$

2 radici imm. in  $\omega$   
 2 radici  $|z_1| = |z_2| = 1$  STAB  
 Scud.

$-\frac{1}{2} < \alpha < 1$   
 2 poli reali  
 $\Downarrow$   
 AS. STAB.

$-\frac{1}{2} < \alpha < 1 \iff$  es. stabil

$\alpha = -\frac{1}{2} \iff$  Deb. neutral

$\left. \begin{array}{l} \alpha < -\frac{1}{2} \\ \alpha > 1 \end{array} \right\} \iff$  instabil

⑥  $\alpha = \frac{1}{2} \implies$  es. stabil

$$u(k) = x(k) - x(k-1) + x(k-2) - x(k-3)$$

$$u(k) = i(k) - i(k-1) + i(k-2) - i(k-3)$$

